MASSIVE MIMO: FUNDAMENTALS AND SYSTEM ISSUES

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Lund Circuit Design Workshop

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You can always lay down more optical fiber; you can never lay down more spectrum!

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SPECTRUM BELOW 5 GHZ:
THE MOST VALUABLE RESOURCE IN THE WORLD!

- FCC AWS-3 spectrum auction, January 2015
  - 65 MHz: 1695-1710 MHz, 1755-1780 MHz, 2155-2180 MHz
  - $41.3 billion

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SPECTRUM BELOW 5 GHZ: THE MOST VALUABLE RESOURCE IN THE WORLD!

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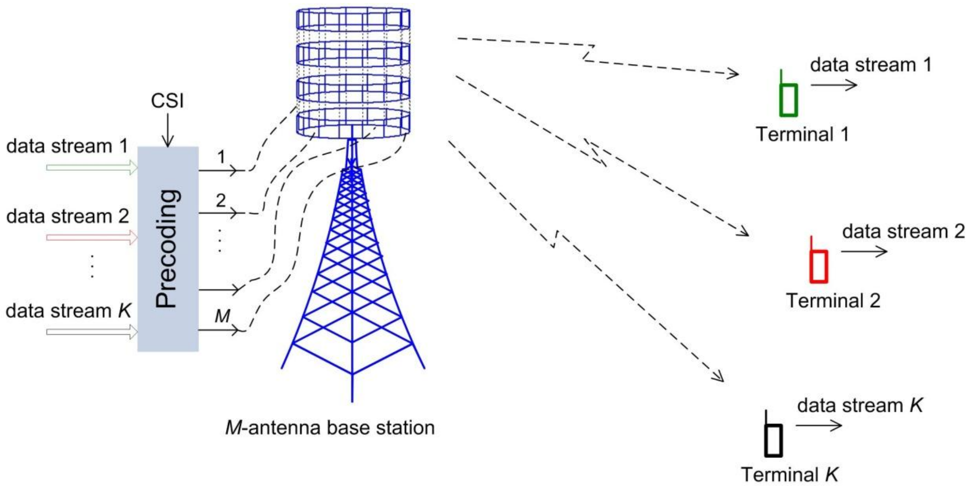
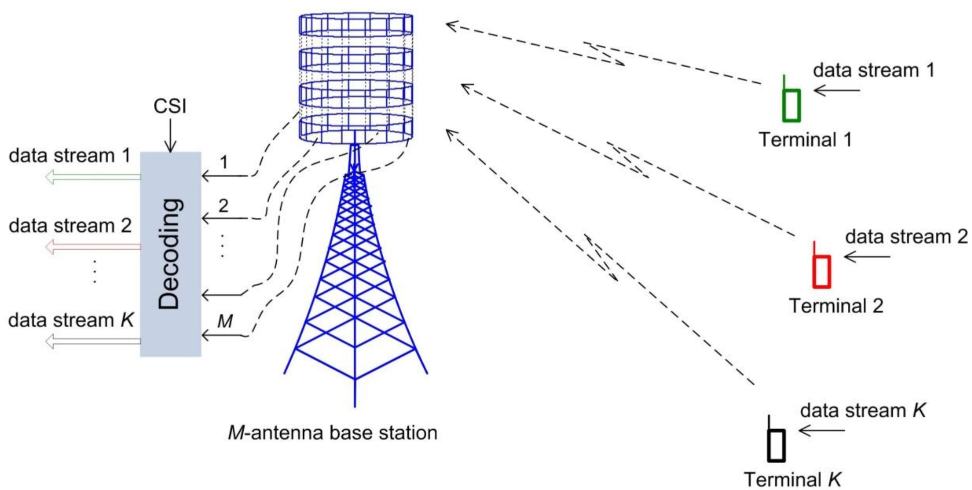
  - 65 MHz: 1695-1710 MHz, 1755-1780 MHz, 2155-2180 MHz
  - $41.3 billion
- *FCC Commissioner Jessica Rosenworcel, October 2, 2014:*
- "What if we issued a challenge in Washington? … Imagine that we decided to reward the first person who finds a way to make spectrum use below 5 GHz 50 or 100 times more efficient over the next decade. The reward could be something simple—say 10 megahertz of spectrum suitable for mobile broadband."

Only advancements in the physical layer can meet this challenge

SPATIAL MULTIPLEXING PUSHED TO AN EXTREME





Massive MIMO serves all users over the *same* time/frequency resources

WHAT IS MASSIVE MIMO?

- Essentials
  - many physically small, low power antennas
  - aggressive spatial multiplexing
  - utilize *measured* channels
- Benefits
  - scalability
  - spectral efficiency
  - simplicity
  - great service to *all* users
  - energy efficiency

Massive MIMO is a game-changer

OUTLINE

- Information theoretic evolution of MIMO
- Science of Massive MIMO
- Case study
  - Optimum pilot re-use
  - Maximum-ratio vs. zero-forcing

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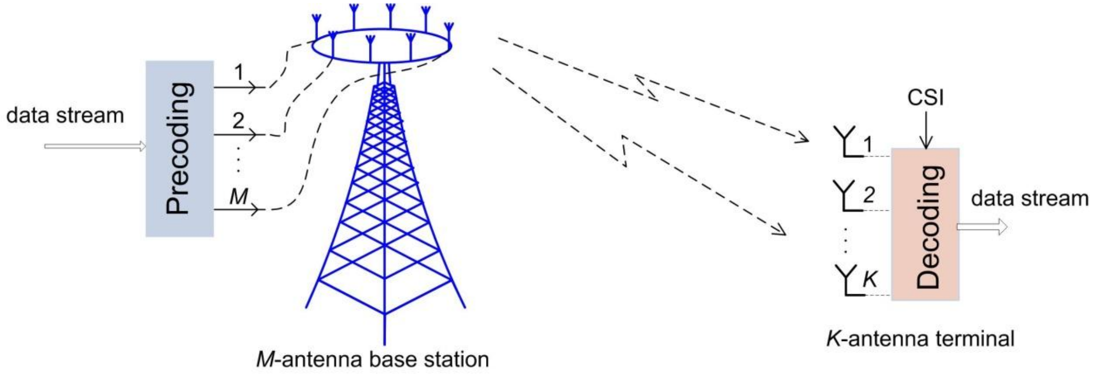
POINT-TO-POINT MIMO

PAULRAJ & KAILATH (1993); FOSCHINI (1995); RALEIGH & CIOFFI (1998)



- Brilliant invention
- But not scalable

  - unfavorable propagation
  - time required for training grows with system size
  - disappointing multiplexing gains at cell edges
  - 8x4 link, -3.0 dB SNR:

# base station antennas	1	2	4	8
bits/second/Hz	1.51	1.83	2.06	2.19

It's critically important to give uniformly good service throughout the cell

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MULTI-USER MIMO

CAIRE & SHAMAI (2003); VISWANATH & TSE (2003); VISHWANATH, JINDAL, & GOLDSMITH (2003)

- Splitting the multi-antenna user into autonomous single-antenna users doesn't decrease the sum-throughput!
- Only single-antenna terminals required
- Propagation is almost always favorable
- But not scalable in its original form
  - dirty-paper coding/decoding needed
  - both ends of link have to know channel

A triumph of Shannon information theory, but not really practical as is

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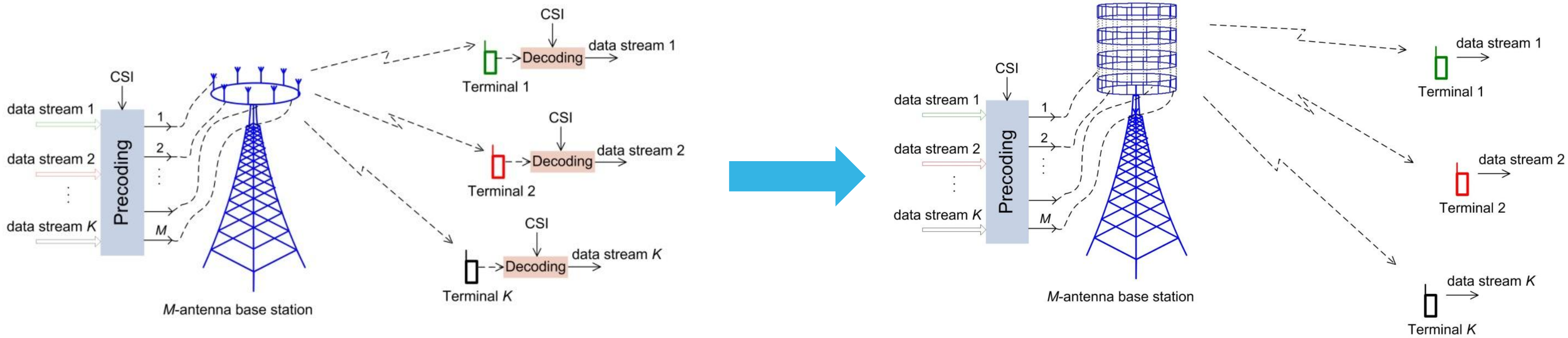
MASSIVE MIMO

MARZETTA (2006); MARZETTA (2010)



- Add many more base station antennas
- Ignore the dictates of Shannon theory
  - channel state information (CSI) only available to the base station
  - use linear pre-coding/de-coding instead of dirty-paper
  - users don’t do any signal processing

The large number of antennas paradoxically makes the problem simpler

MASSIVE MIMO: MORE THAN JUST MANY ANTENNAS

- Using *measured* channels: Beamforming gain grows linearly with number of antennas, irrespective of the noisiness of the measurements
- Frequency-independent power control: Based solely on long-scale (slow) fading, is exceedingly effective
- Pilot contamination: Ultimate limitation in non-cooperative multi-cell systems

No new mathematics, but a new philosophy!

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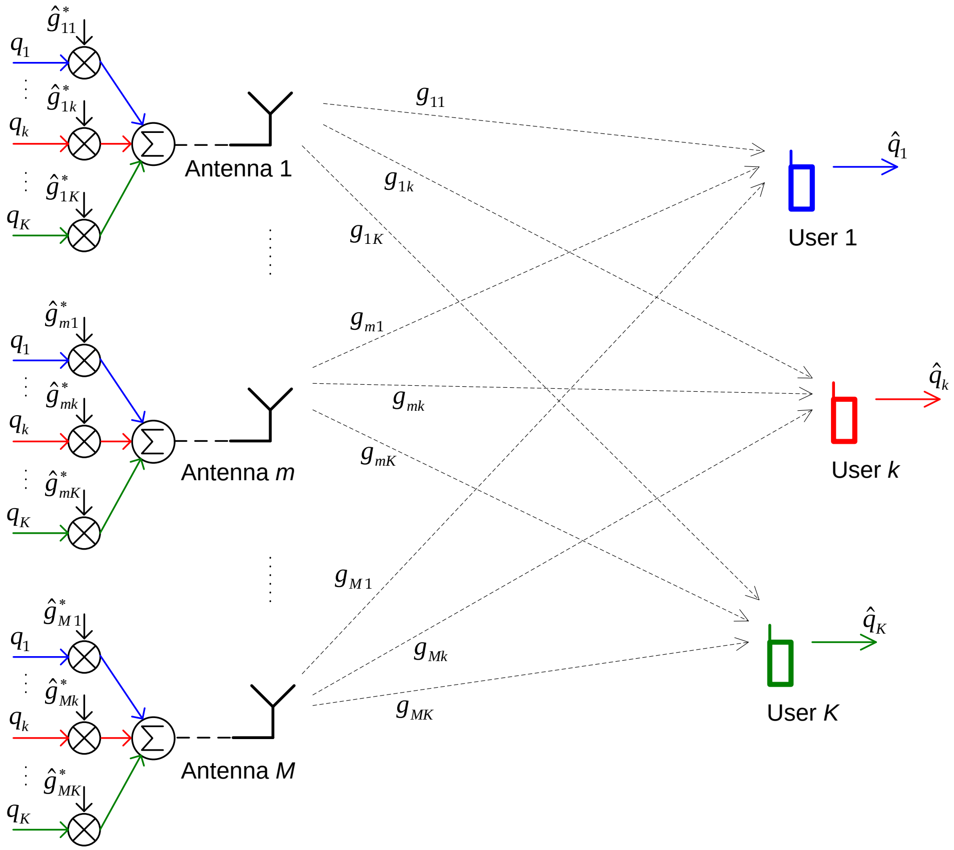
DOWNLINK DATA TRANSMISSION: CONJUGATE BEAMFORMING

ANTENNAS TRANSMIT THE WEIGHTED MESSAGE-BEARING SYMBOLS TO ARRIVE IN-PHASE AT THE INTENDED USER & OUT-OF-PHASE ELSEWHERE

- Information-bearing symbols combined with *measured* channel characteristics to create transmitted signals
- Decentralized array architecture



The simplest possible pre-coding, but often very effective

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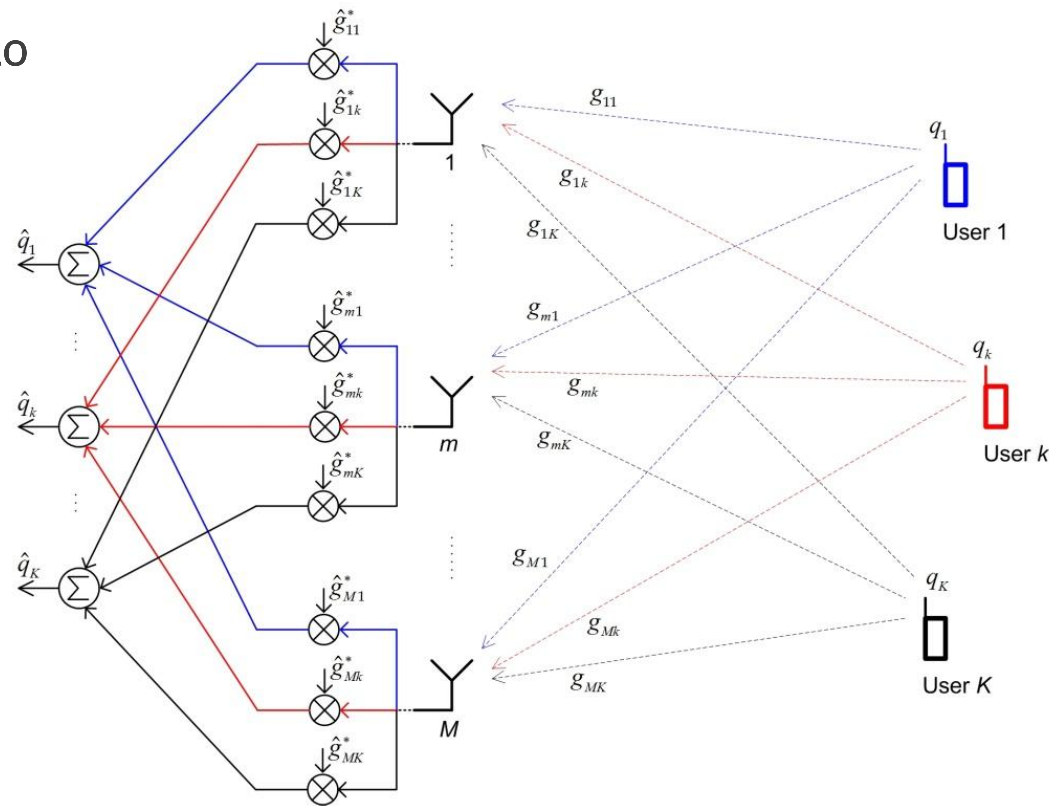
UPLINK DATA TRANSMISSION: MATCHED FILTERING

BASE STATION WEIGHTS AND ADDS RECEIVED SIGNALS FOR CONSTRUCTIVE REINFORCEMENT OF THE TRANSMISSION FROM EACH USER

- Properties and advantages similar to conjugate beamforming



For high SINRs, zero-forcing may outperform conjugate beamforming/matched-filtering

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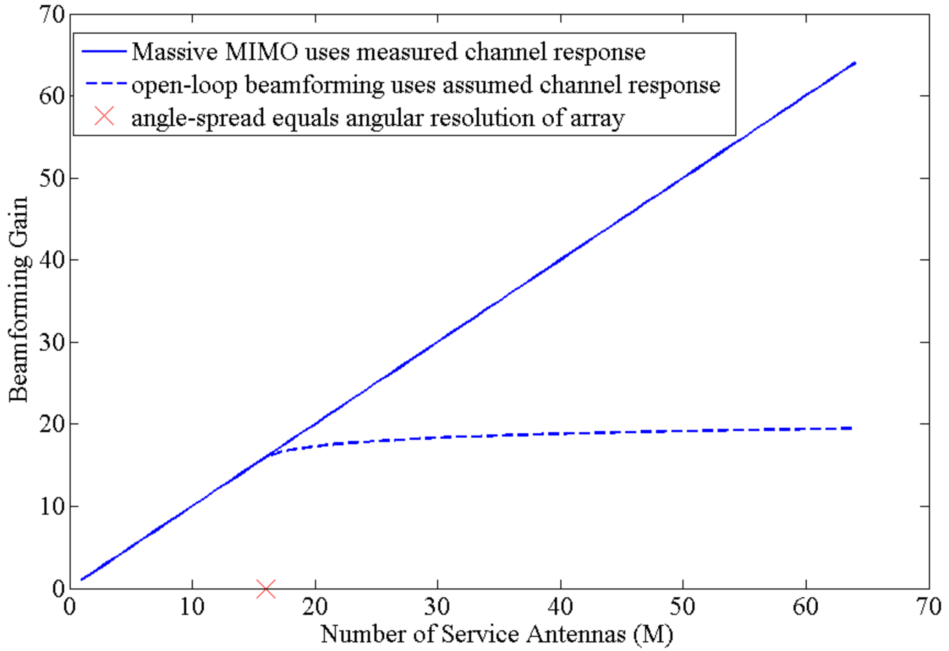
WHY SO IMPORTANT TO DO BEAMFORMING WITH *MEASURED* PROPAGATION?

- *Measured* channels
    - scalable
    - gain grows linearly with number of antennas
        - irrespective of noisiness of CSI
        - no tightening of array tolerance required
- *Assumed* channels
    - not scalable
    - gain eventually grows only logarithmically



If open-loop beamforming, then not Massive MIMO!

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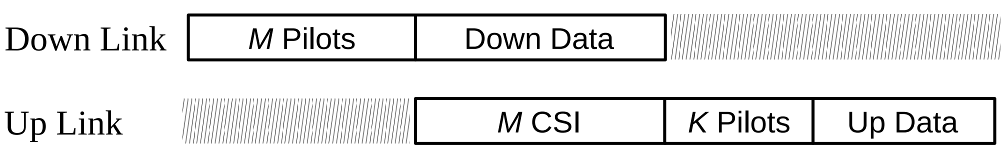
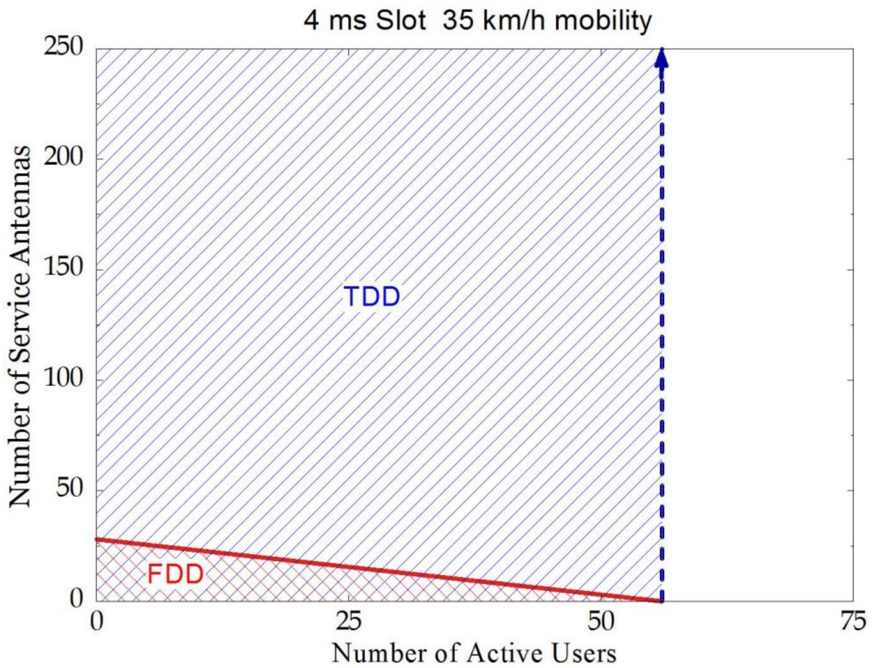
TDD SLOT STRUCTURE ENSURES TIMELY CHANNEL-STATE INFORMATION: M SERVICE-ANTENNAS, K USERS

- TDD slot: training time $\propto K$

- FDD slot: training time $\propto 2M + K$





Mobility limits the number of active users; FDD is a disaster!

PILOT CONTAMINATION

- For mobile users, there is a limited number of orthogonal pilots
- When the same pilot is transmitted by more than one user:
  - base station obtains a *linear combination* of channels
    - extra pilot power doesn't help
  - coherent interference
    - doesn't disappear with more antennas

Pilot contamination has always existed, but was never noticed!

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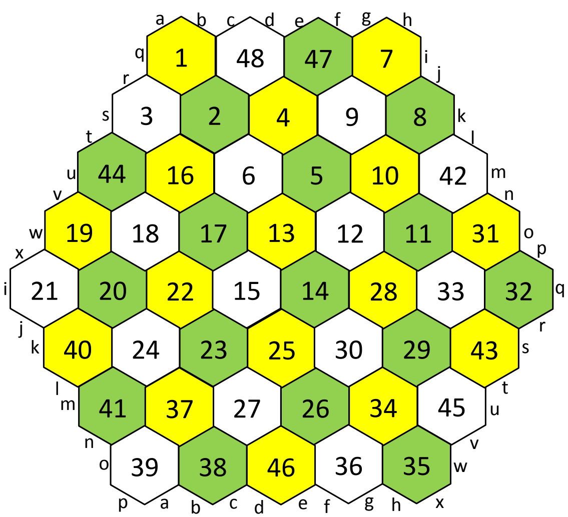
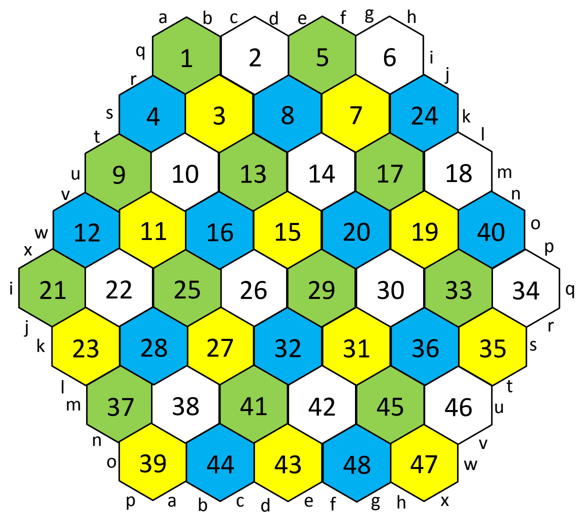
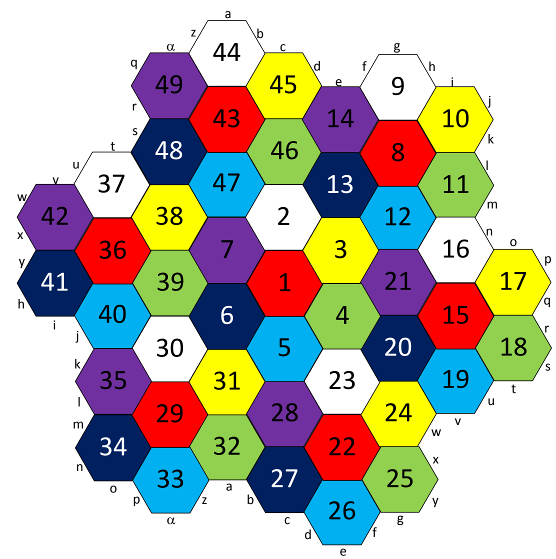
PILOT RE-USE FACTOR 3, 4, 7: PUSH CONTAMINATING CELLS FARTHER AWAY FROM HOME CELL







The cost: extra overhead

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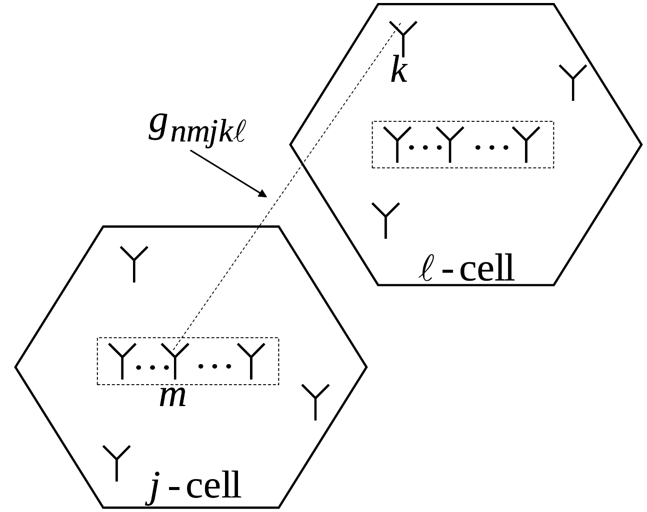
•Propagation Model

$$g_{mk} = \underbrace{\beta_k^{1/2}}_{\text{slow}} \cdot \underbrace{h_{mk}}_{\text{fast}}, \quad m: \text{ antenna} \quad k: \text{ terminal}$$



- Slow fading and fast fading
- Slow fading comprises geometric attenuation (Hata Model) combined with log-normal shadow fading
  - Constant with respect to frequency and service antenna
  - Easy to estimate
  - Assumed known a-priori
- Fast fading
  - Rayleigh CN(0,1), iid with respect to antenna, terminal
  - Piecewise constant and iid from one frequency smoothness interval to another
  - Unknown a-priori
  - Estimated from up link pilots & TDD reciprocity

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DOWN LINK DATA: LINEAR PRE-CODING

$$\underbrace{\mathbf{s}}_{M\times 1} = \underbrace{\mathbf{A}}_{M\times K} \underbrace{\mathbf{D}_{\boldsymbol{\eta}}^{1/2}}_{K\times 1} \underbrace{\mathbf{q}}_{K\times 1}$$

$\{q_k\}$: iid, CN(0,1) $\quad E\{\mathbf{s}^{\mathrm{H}}\mathbf{s}\}=1 \quad$ power - control : $\boldsymbol{\eta} \geq \mathbf{0},\ \mathbf{1}^{\mathrm{T}}\boldsymbol{\eta}=1$

- Conjugate beam-forming

$$\mathbf{A} = \frac{1}{\sqrt{M}}\hat{\mathbf{G}}^{*}\mathbf{D}_{\boldsymbol{\gamma}}^{-1/2}$$

  - permits de-centralized architecture and processing

- Zero-forcing

$$\mathbf{A} = \sqrt{M-K}\,\hat{\mathbf{G}}^{*}\left(\hat{\mathbf{G}}^{\mathrm{T}}\hat{\mathbf{G}}^{*}\right)^{-1}\mathbf{D}_{\boldsymbol{\gamma}}^{1/2}$$

  - Implementing linear pre-coding takes more computations than QR factorization!

DOWN LINK: CONJUGATE BEAM-FORMING

$$\mathbf{x} = \sqrt{\frac{\rho_{\mathrm{f}}}{M}}\mathbf{G}^{\mathrm{T}}\hat{\mathbf{G}}^{*}\mathbf{D}_{\boldsymbol{\gamma}}^{-1/2}\mathbf{D}_{\boldsymbol{\eta}}^{1/2}\mathbf{q} + \mathbf{w} = \sqrt{\frac{\rho_{\mathrm{f}}}{M}}\hat{\mathbf{G}}^{\mathrm{T}}\hat{\mathbf{G}}^{*}\mathbf{D}_{\boldsymbol{\gamma}}^{-1/2}\mathbf{D}_{\boldsymbol{\eta}}^{1/2}\mathbf{q} + \mathbf{w} - \sqrt{\frac{\rho_{\mathrm{f}}}{M}}\tilde{\mathbf{G}}^{\mathrm{T}}\hat{\mathbf{G}}^{*}\mathbf{D}_{\boldsymbol{\gamma}}^{-1/2}\mathbf{D}_{\boldsymbol{\eta}}^{1/2}\mathbf{q}$$

$$x_k = \sqrt{\frac{\rho_{\mathrm{f}}\eta_k}{M\gamma_k}}\hat{\mathbf{g}}_k^{\mathrm{T}}\hat{\mathbf{g}}_k^{*}q_k + w_k - \sqrt{\frac{\rho_{\mathrm{f}}}{M}}\tilde{\mathbf{g}}_k^{\mathrm{T}}\hat{\mathbf{G}}^{*}\mathbf{D}_{\boldsymbol{\gamma}}^{-1/2}\mathbf{D}_{\boldsymbol{\eta}}^{1/2}\mathbf{q} + \hat{\mathbf{g}}_k^{\mathrm{T}} \cdot \sum_{n\neq k}\sqrt{\frac{\rho_{\mathrm{f}}\eta_n}{M\gamma_n}}\hat{\mathbf{g}}_n^{*}q_n$$

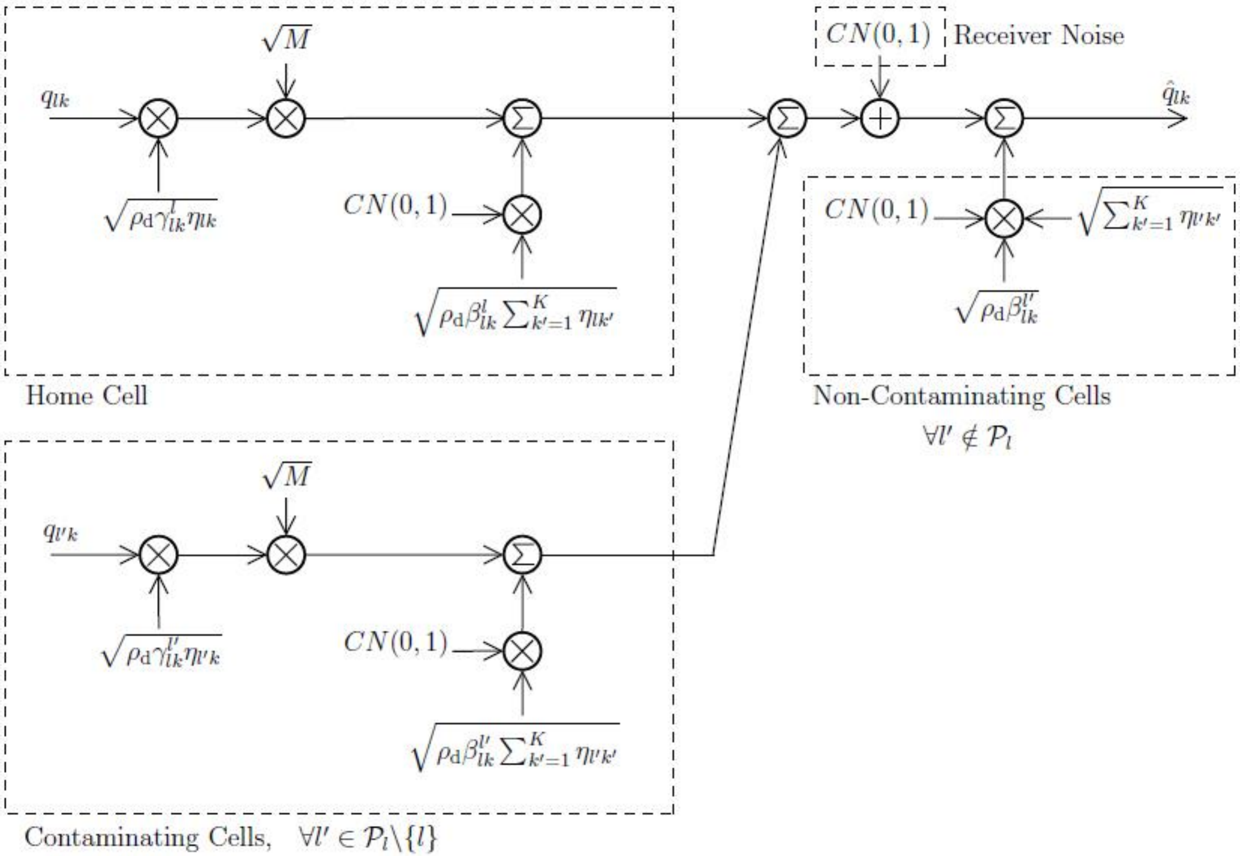
$$= \underbrace{\sqrt{\frac{\rho_{\mathrm{f}}\eta_k}{M\gamma_k}}\mathrm{E}\left\{\hat{\mathbf{g}}_k^{\mathrm{T}}\hat{\mathbf{g}}_k^{*}\right\}q_k}_{(0)} + \underbrace{w_k}_{(1)} - \underbrace{\sqrt{\frac{\rho_{\mathrm{f}}}{M}}\tilde{\mathbf{g}}_k^{\mathrm{T}}\hat{\mathbf{G}}^{*}\mathbf{D}_{\boldsymbol{\gamma}}^{-1/2}\mathbf{D}_{\boldsymbol{\eta}}^{1/2}\mathbf{q}}_{(2)}$$

$$+ \underbrace{\hat{\mathbf{g}}_k^{\mathrm{T}} \cdot \sum_{n\neq k}\sqrt{\frac{\rho_{\mathrm{f}}\eta_n}{M\gamma_n}}\hat{\mathbf{g}}_n^{*}q_n}_{(3)} + \underbrace{\sqrt{\frac{\rho_{\mathrm{f}}\eta_k}{M\gamma_k}}\left(\hat{\mathbf{g}}_k^{\mathrm{T}}\hat{\mathbf{g}}_k^{*} - \mathrm{E}\left\{\hat{\mathbf{g}}_k^{\mathrm{T}}\hat{\mathbf{g}}_k^{*}\right\}\right)q_k}_{(4)}$$

(0) desired signal	(1) receiver noise	(2) channel estimation error	(3) channel non-orthogonality	(4) beam-forming gain uncertainty
$M\rho_{\mathrm{f}}\eta_k\gamma_k$	1	$\rho_{\mathrm{f}}(\beta_k - \gamma_k)$	$\rho_{\mathrm{f}}\gamma_k\sum_{n\neq k}\eta_n$	$\rho_{\mathrm{f}}\gamma_k\eta_k$

$$\mathrm{SINR}_k = \frac{\eta_k M\rho_{\mathrm{f}}\gamma_k}{1+\rho_{\mathrm{f}}(\beta_k - \gamma_k) + \rho_{\mathrm{f}}\gamma_k\sum_{n\neq k}\eta_n + \rho_{\mathrm{f}}\gamma_k\eta_k} = \frac{\eta_k M\rho_{\mathrm{f}}\gamma_k}{1+\rho_{\mathrm{f}}\left(\sum_{n=1}^{K}\eta_n\right)\beta_k}$$

DOWNLINK MAXIMUM-RATIO: EFFECTIVE CHANNEL



$$\text{SINR}_{\ell k} = \frac{M\rho_{\text{f}}\eta_{k\ell}\gamma_{\ell k\ell}}{1 + \sum_{j \in \text{all}} \rho_{\text{f}} \left(\sum_{n=1}^{K} \eta_{nj} \right) \beta_{jk\ell} + \sum_{j \in \text{PC}_\ell} M\rho_{\text{f}}\eta_{kj}\gamma_{jk\ell}}, \quad \beta: \text{m.s. channel}, \gamma: \text{m.s. estimate}$$

Massive MIMO creates a flat channel to each terminal

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SINR FOR K-TH TERMINAL IN ELL-TH CELL

β: m.s. channel $\quad \gamma$: m.s. estimate $\quad \beta-\gamma$: m.s. error $\quad \eta$: power control

	Conjugate Beam Forming / Matched Filtering	Zero Forcing
Down Link	$\dfrac{M\rho_{\mathrm{f}}\eta_{k\ell}\gamma_{\ell k\ell}}{1+\sum_{j\in\mathrm{all}}\rho_{\mathrm{f}}\left(\sum_{n=1}^{K}\eta_{nj}\right)\beta_{jk\ell}+\sum_{j\in\mathrm{PC}_\ell}M\rho_{\mathrm{f}}\eta_{kj}\gamma_{jk\ell}}$	$\dfrac{(M-K)\rho_{\mathrm{f}}\eta_{k\ell}\gamma_{\ell k\ell}}{1+\sum_{j\notin\ell\cup\mathrm{PC}_\ell}\rho_{\mathrm{f}}\left(\sum_{n=1}^{K}\eta_{nj}\right)\beta_{jk\ell}+\sum_{j\in\ell\cup\mathrm{PC}_\ell}\rho_{\mathrm{f}}\left(\sum_{n=1}^{K}\eta_{nj}\right)(\beta_{jk\ell}-\gamma_{jk\ell})+\sum_{j\in\mathrm{PC}_\ell}(M-K)\rho_{\mathrm{f}}\eta_{kj}\gamma_{jk\ell}}$
Up Link	$\dfrac{M\rho_{\mathrm{r}}\eta_{k\ell}\gamma_{\ell k\ell}}{1+\sum_{j\in\mathrm{all}}\rho_{\mathrm{r}}\sum_{n=1}^{K}\eta_{nj}\beta_{\ell nj}+\sum_{j\in\mathrm{PC}_\ell}M\rho_{\mathrm{r}}\eta_{kj}\gamma_{\ell kj}}$	$\dfrac{(M-K)\rho_{\mathrm{r}}\eta_{k\ell}\gamma_{\ell k\ell}}{1+\sum_{j\notin\ell\cup\mathrm{PC}_\ell}\rho_{\mathrm{r}}\sum_{n=1}^{K}\eta_{nj}\beta_{\ell nj}+\sum_{j\in\ell\cup\mathrm{PC}_\ell}\rho_{\mathrm{r}}\sum_{n=1}^{K}\eta_{nj}(\beta_{\ell nj}-\gamma_{\ell nj})+\sum_{j\in\mathrm{PC}_\ell}(M-K)\rho_{\mathrm{r}}\eta_{kj}\gamma_{\ell kj}}$

- Inequality constraints on SINR equivalent to linear inequality constraints on power control variables

CASE STUDIES: OPTIMUM PILOT RE-USE FACTOR; MAXIMUM-RATIO VS. ZERO-FORCING

	Dense Urban	Suburban
Carrier frequency(GHz)	1.9	1.9
TDD spectral bandwidth (MHz)	20	20
Slot duration (ms)	2	1
User allowed mobility (km/h)	71	142
Uplink radiated power/user (mW)	200	200
Number of service antennas	64	256
Total downlink radiated power (W)	1	1
Active users/cell	18	18
Cell radius (km)	.50	2.0
Power control	Max/min	Max/min
95% likely throughput/terminal Mb/s zero-forcing	4.1 down, 2.6 up	3.1 down, 1.1 up
95% likely throughput/terminal Mb/s maximum-ratio	4.5 down, 3.1 up	3.2 down, 1.1 up

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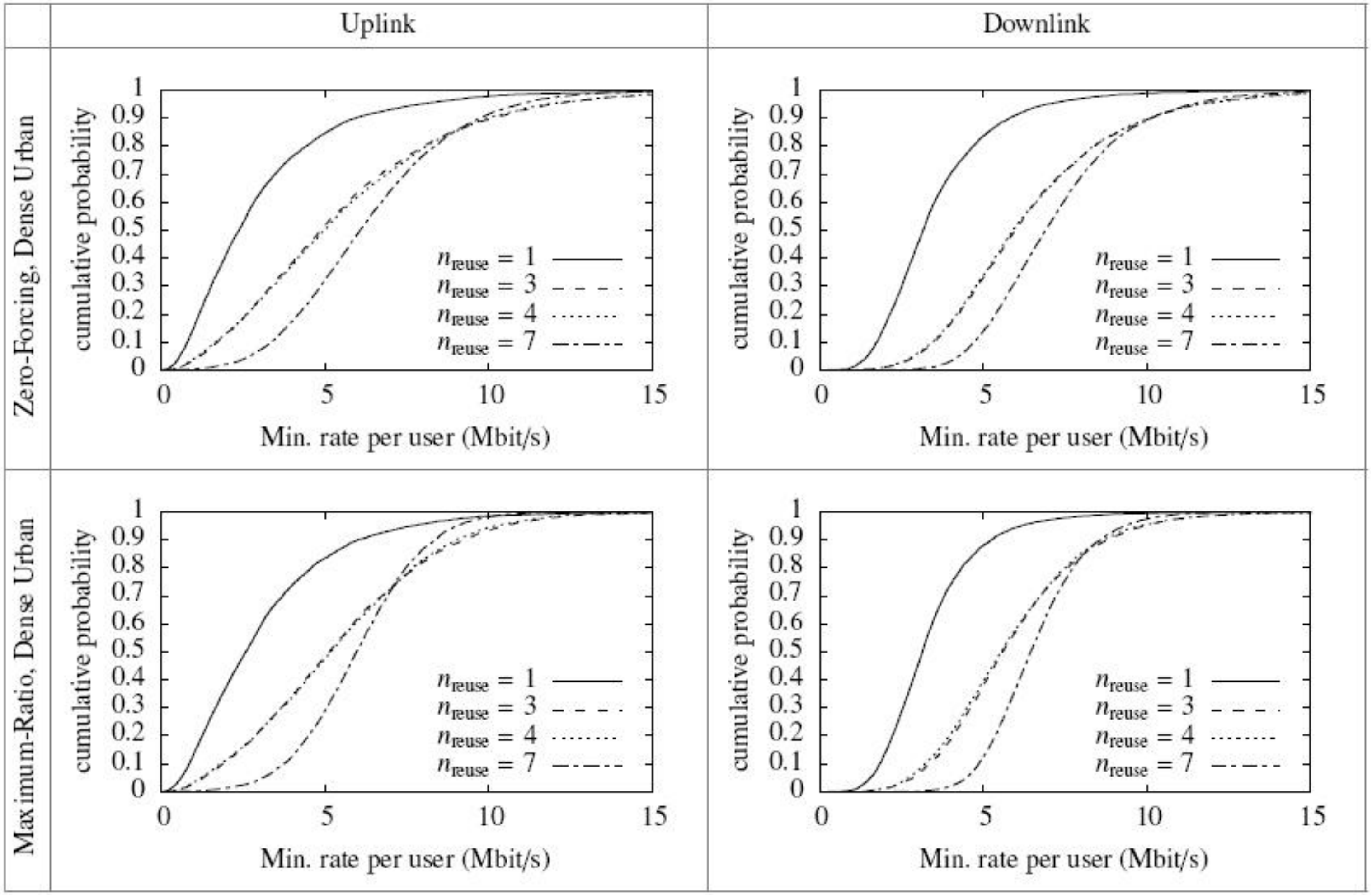
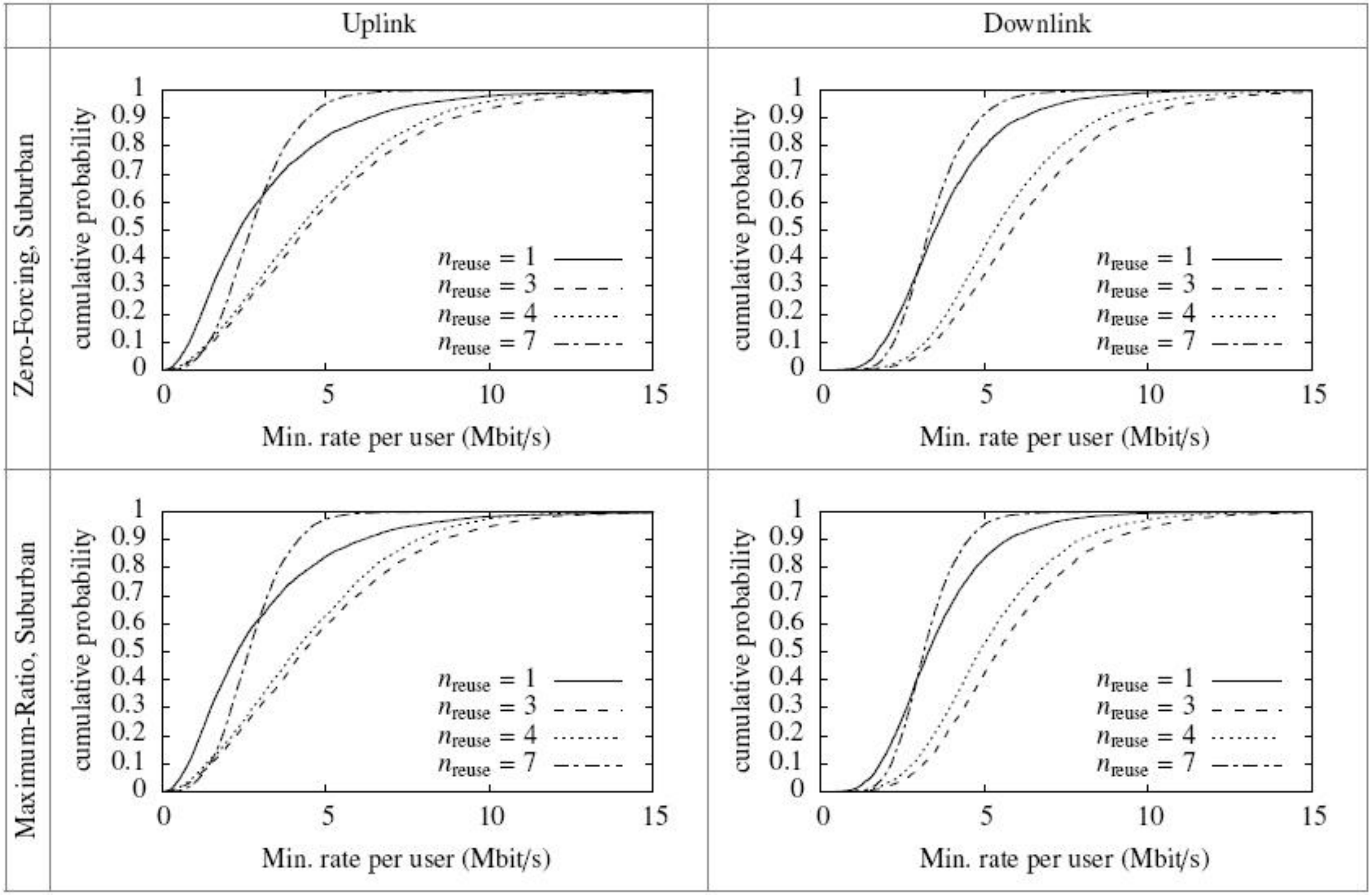
NET THROUGHPUT PER USER: DENSE URBAN



Maximum-ratio + pilot re-use 7 best for Dense Urban

NET THROUGHPUT PER USER: SUBURBAN



Maximum-ratio + pilot re-use 3 best for Suburban

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NON-CELLULAR MASSIVE MIMO

- Backhaul for small-cells
- Cell-Free Massive MIMO
  - M randomly distributed access points serve K users over an entire city
- Fixed wireless access to homes
- Multicasting
  - Deliberately create and take advantage of pilot contamination!

MASSIVE MIMO IN NON-ELECTROMAGNETIC MEDIA

- Hyperbolic
  - acoustic waves
  - elastic waves
- Parabolic
  - heat diffusion
- Eliptical
  - electric current: *Ground Telegraphy* (Richard Courant, Arnold Sommerfeld, Lee de Forest)

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MASSIVE SENSOR TELEMETRY

- Outstanding examples of signal processing
  - 3D reflection seismology
  - Computer tomography
  - Synthetic aperture radar
- Essential to sample data spatially and temporally at Nyquist rate
  - Proper sampling and huge data sets make physics-based signal processing easier!
- Uplink Massive MIMO transports data intact and in real time

Massive MIMO means more than entertainment enablement

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